## **Direct AC/AC Matrix Converters**



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- have received considerable attention in recent years
- may become a good alternative to PWM-VSI topology
- Bi-directional power flow
- Sinusoidal input/output waveforms
- Controllable input power factor
- Compact design, due to the lack of dc-link capacitors for energy storage

**Topology complexity of the matrix converter makes study a hard task** 



Basic scheme of matrix converters.

#### Space Vector Modulation (SVM) algorithm

#### SVM completely exploit the possibility of matrix converters to

- control the input power factor regardless the output power factor
- fully utilize the input voltages
- reduce the number of switch commutations in each cycle period.

# SVM allows an immediate comprehension of the modulation process

- without the need for a fictitious DC link
- avoiding the addition of the third harmonic components.



#### Switching Configurations Employed

Switching configuration list	Switches On			vo	α <sub>0</sub>	i <sub>i</sub>	β <sub>i</sub>				
+1	$S_{11}$	$S_{22}$	$S_{32}$	$2/3 v_{12i}$	0	$2/\sqrt{3} i_{ol}$	-π⁄6				
-1	$S_{12}$	$S_{21}$	$S_{31}$	$-2/3 v_{12i}$	0	$-2/\sqrt{3} i_{ol}$	- <i>π</i> /6				
+2	$S_{12}$	$S_{23}$	$S_{33}$	$2/3 v_{23i}$	0	$2/\sqrt{3} i_{ol}$	$\pi/2$				
-2	$S_{13}$	$S_{22}$	$S_{32}$	$-2/3 v_{23i}$	0	$-2/\sqrt{3} i_{ol}$	$\pi/2$				
+3	$S_{13}$	$S_{21}$	$S_{31}$	$2/3 v_{31i}$	0	$2/\sqrt{3} i_{ol}$	7π/6				
-3	$S_{11}$	$S_{23}$	$S_{33}$	$-2/3 v_{31i}$	0	$-2/\sqrt{3} i_{o1}$	7π/6				
+4	$S_{12}$	$S_{21}$	$S_{32}$	$2/3 v_{12i}$	2π/3	$2/\sqrt{3} i_{o2}$	- <i>π</i> /6				
-4	$S_{11}$	$S_{22}$	$S_{31}$	$-2/3 v_{12i}$	2π/3	$-2/\sqrt{3} i_{o2}$	- <i>π</i> /6				
+5	$S_{13}$	$S_{22}$	$S_{33}$			Garage					
-5	$S_{12}$	$S_{23}$	$S_{32}$	Active configurations							
+6	$S_{11}$	$S_{23}$	$S_{31}$	2, 5-v <sub>31i</sub>	2105	2/ vJ 1 <sub>02</sub>	1100				
-6	$S_{13}$	$S_{21}$	$S_{33}$	<i>-2/3 v<sub>31i</sub></i>	2π/3	$-2/\sqrt{3} i_{o2}$	7π/6				
+7	$S_{12}$	$S_{22}$	$S_{31}$	$2/3 v_{12i}$	4π/3	$2/\sqrt{3} i_{o3}$	- <i>π</i> /6				
-7	$S_{11}$	$S_{21}$	$S_{32}$	$-2/3 v_{12i}$	$4\pi/3$	$-2/\sqrt{3} i_{o3}$	- <i>π</i> /6				
+8	$S_{13}$	$S_{23}$	$S_{32}$	$2/3 v_{23i}$	$4\pi/3$	$2/\sqrt{3} i_{o3}$	$\pi/2$				
-8	$S_{12}$	$S_{22}$	$S_{33}$	-2/3 v <sub>23i</sub>	$4\pi/3$	-2/ √3 i <sub>03</sub>	$\pi/2$				
+9	$S_{11}$	$S_{21}$	$S_{33}$	2/3 v <sub>31i</sub>	$4\pi/3$	$2/\sqrt{3} i_{o3}$	7π/6				
-9	$S_{13}$	$S_{23}$	$S_{31}$	<i>-2/3 v<sub>31i</sub></i>	4π/3	$-2/\sqrt{3} i_{o3}$	7π/6				
$0_1$ $0_2$	$\frac{S_{11}}{S_{12}}$	$\overline{\begin{array}{c}S_{21}\\S_{22}\\\widetilde{\end{array}}}$	$\frac{S_{31}}{S_{32}}$	Zero configurations							

#### **Output Voltage and Input Current Space Vectors**



Sectors and directions of the output line-to-neutral voltage vectors.



Sectors and directions of the input current vectors.

At any sampling instant  $\bar{v}_o$  and  $\phi_i$  are known (reference quantities)

 $\overline{v}_i$  is imposed by the source (known by measurements)



**SVM Algorithm** is based on the selection of <u>4 active</u> <u>configurations</u> that are applied for suitable time intervals within each cycle period  $T_c$ .

The <u>zero configurations</u> are applied to complete  $T_c$ .

The control of  $\varphi_i$  can be achieved by controlling the phase angle  $\beta_i$  of the input current vector.

	Sector of the output voltage vector $(K_v)$												
		1 or 4				2 or 5			3 or 6				
he ent	1 or 4	+9	+7	+3	+1	+6	+4	+9	+7	+3	+1	+6	+4
or of t it curre tor (K	2 or 5	+8	+9	+2	+3	+5	+6	+8	+9	+2	+3	+5	+6
Sect inpu	3 or 6	+7	+8	+1	+2	+4	+5	+7	+8	+1	+2	+4	+5
		Ι	Π	III	IV	Ι	II	III	IV	Ι	Π	III	IV

Selection of the active switching configurations for each combination of

- output voltage sector
- input current sector

## **Basic equations of the SVM algorithm**

$$\begin{split} \overline{v}_{o}^{'} &= \overline{v}_{o}^{I} \delta^{I} + \overline{v}_{o}^{II} \delta^{II} = \frac{2}{\sqrt{3}} v_{o} \cos(\widetilde{\alpha}_{o} - \frac{\pi}{3}) e^{j[(K_{v} - 1)\pi/3 + \pi/3]} \\ \overline{v}_{o}^{''} &= \overline{v}_{o}^{III} \delta^{III} + \overline{v}_{o}^{IV} \delta^{IV} = \frac{2}{\sqrt{3}} v_{o} \cos(\widetilde{\alpha}_{o} + \frac{\pi}{3}) e^{j[(K_{v} - 1)\pi/3]} \\ \left(\overline{i}_{i}^{I} \delta^{I} + \overline{i}_{i}^{II} \delta^{II}\right) \cdot j e^{j\widetilde{\beta}_{i}} e^{j(K_{i} - 1)\pi/3} = 0 \\ \left(\overline{i}_{i}^{III} \delta^{III} + \overline{i}_{i}^{IV} \delta^{IV}\right) \cdot j e^{j\widetilde{\beta}_{i}} e^{j(K_{i} - 1)\pi/3} = 0 \end{split}$$

To satisfy to the requirements of the

- reference output voltage vector
- input current displacement angle.

### **Solutions of the Basic Equations**

$$\begin{split} \delta^{I} &= (-1)^{K_{v}+K_{i}} \frac{2}{\sqrt{3}} q \frac{\cos(\tilde{\alpha}_{o}-\pi/3)\cos(\tilde{\beta}_{i}-\pi/3)}{\cos\varphi_{i}} \\ \delta^{II} &= (-1)^{K_{v}+K_{i}+1} \frac{2}{\sqrt{3}} q \frac{\cos(\tilde{\alpha}_{o}-\pi/3)\cos(\tilde{\beta}_{i}+\pi/3)}{\cos\varphi_{i}} \\ \delta^{III} &= (-1)^{K_{v}+K_{i}+1} \frac{2}{\sqrt{3}} q \frac{\cos(\tilde{\alpha}_{o}+\pi/3)\cos(\tilde{\beta}_{i}-\pi/3)}{\cos\varphi_{i}} \\ \delta^{IV} &= (-1)^{K_{v}+K_{i}} \frac{2}{\sqrt{3}} q \frac{\cos(\tilde{\alpha}_{o}+\pi/3)\cos(\tilde{\beta}_{i}+\pi/3)}{\cos\varphi_{i}} \end{split}$$

Applied for any combination of > output voltage sector  $K_v$ > input current sector  $K_i$ 

#### For the feasibility of the control strategy.

$$\left|\delta^{I}\right| + \left|\delta^{II}\right| + \left|\delta^{III}\right| + \left|\delta^{IV}\right| \le 1$$

Otherwise: the instantaneous values of the input voltages do not allow to satisfy the requirements of output voltage and input power factor.